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INA128UA/2K5

TI, Texas Instruments

Instrumentation Amplifiers Precision Low Power Instrumentation
Amp

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INA12x Precision, Low-Power Instrumentation Amplifiers

A newer version of this device is now available: INA828

1 Features

- A newer version of this device is now available: [INA828](#)
- Low offset voltage: 50 μV maximum
- Low drift: 0.5 $\mu\text{V}/^\circ\text{C}$ maximum
- Low Input Bias Current: 5 nA maximum
- High CMR: 120 dB minimum
- Inputs protected to $\pm 40\text{ V}$
- Wide supply range: $\pm 2.25\text{ V}$ to $\pm 18\text{ V}$
- Low quiescent current: 700 μA
- Packages: 8-pin plastic DIP, SO-8

2 Applications

- Bridge amplifier
- Thermocouple amplifier
- RTD sensor amplifier
- Medical instrumentation
- Data acquisition

3 Description

The INA128 and INA129 are low-power, general purpose instrumentation amplifiers offering excellent accuracy. The versatile 3-op amp design and small size make these amplifiers ideal for a wide range of applications. Current-feedback input circuitry provides wide bandwidth even at high gain (200 kHz at $G = 100$).

A single external resistor sets any gain from 1 to 10,000. The INA128 provides an industry-standard gain equation; the INA129 gain equation is compatible with the AD620.

The INA12x is available in 8-pin plastic DIP and SO-8 surface-mount packages, specified for the -40°C to $+85^\circ\text{C}$ temperature range. The INA128 is also available in a dual configuration, the INA2128.

The upgraded [INA828](#) offers a lower input bias current (0.6 nA maximum) and lower noise (7 $\text{nV}/\sqrt{\text{Hz}}$) at the same quiescent current. See the [Device Comparison Table](#) for a selection of precision instrumentation amplifiers from Texas Instruments.

Device Information⁽¹⁾

PART NUMBER	PACKAGE	BODY SIZE (NOM)
INA128, INA129	SOIC (8)	3.91 mm x 4.90 mm
	PDIP (8)	6.35 mm x 9.81 mm

(1) For all available packages, see the package option addendum at the end of the data sheet.

Simplified Schematic

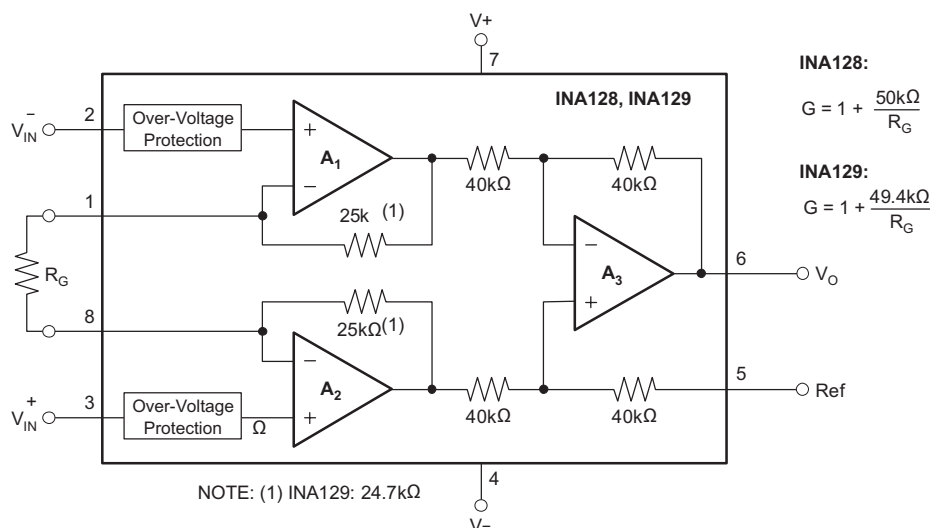


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4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision D (January 2018) to Revision E Page

- Added information about the newer, upgraded INA828 1
- Added *Device Comparison Table* 3

Changes from Revision C (October 2015) to Revision D Page

- Added top navigator icon for TI Reference Design 1
- Changed " $\pm 0.5 \pm 0/G$ " to " $\pm 0.5 \pm 20/G$ " in MAX column of Offset voltage RTI vs temperature row of *Electrical Characteristics*..... 5

Changes from Revision B (February 2005) to Revision C Page

- Added *ESD Ratings* table, *Feature Description* section, *Device Functional Modes*, *Application and Implementation* section, *Power Supply Recommendations* section, *Layout* section, *Device and Documentation Support* section, and *Mechanical, Packaging, and Orderable Information* section. 1

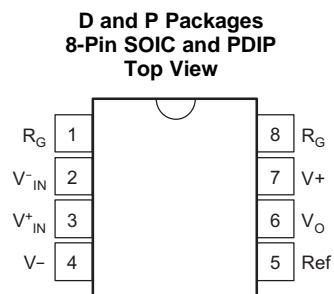
5 Device Comparison Table

DEVICE	DESCRIPTION	GAIN EQUATION	RG PINS AT PIN
INA828	50- μ V Offset, 0.5 μ V/ $^{\circ}$ C V_{OS} drift, 7-nV/ $\sqrt{\text{Hz}}$ Noise, Low-Power, Precision Instrumentation Amplifier	$G = 1 + 50 \text{ k}\Omega / R_G$	1, 8
INA819	35- μ V Offset, 0.4 μ V/ $^{\circ}$ C V_{OS} drift, 8-nV/ $\sqrt{\text{Hz}}$ Noise, Low-Power, Precision Instrumentation Amplifier	$G = 1 + 50 \text{ k}\Omega / R_G$	2, 3
INA821	35- μ V Offset, 0.4 μ V/ $^{\circ}$ C V_{OS} drift, 7-nV/ $\sqrt{\text{Hz}}$ Noise, High-Bandwidth, Precision Instrumentation Amplifier	$G = 1 + 49.4 \text{ k}\Omega / R_G$	2, 3
INA828	50- μ V Offset, 0.5 μ V/ $^{\circ}$ C V_{OS} drift, 7-nV/ $\sqrt{\text{Hz}}$ Noise, Low-Power, Precision Instrumentation Amplifier	$G = 1 + 50 \text{ k}\Omega / R_G$	1, 8
INA333	25- μ V V_{OS} , 0.1 μ V/ $^{\circ}$ C V_{OS} drift, 1.8-V to 5-V, RRO, 50- μ A I_Q , chopper-stabilized INA	$G = 1 + 100 \text{ k}\Omega / R_G$	1, 8
PGA280	20-mV to ± 10 -V programmable gain IA with 3-V or 5-V differential output; analog supply up to ± 18 V	digital programmable	N/A
INA159	$G = 0.2$ V differential amplifier for ± 10 -V to 3-V and 5-V conversion	$G = 0.2 \text{ V/V}$	N/A
PGA112	Precision programmable gain op amp with SPI	digital programmable	N/A

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6 Pin Configuration and Functions**Pin Functions**

PIN		I/O	DESCRIPTION
NAME	NO.		
REF	5	I	Reference input. This pin must be driven by low impedance or connected to ground.
R _G	1,8	—	Gain setting pin. For gains greater than 1, place a gain resistor between pin 1 and pin 8.
V ₋	4	—	Negative supply
V ₊	7	—	Positive supply
V _{IN-}	2	I	Negative input
V _{IN+}	3	I	Positive input
V _O	6	I	Output

7 Specifications**7.1 Absolute Maximum Ratings**over operating free-air temperature range (unless otherwise noted) ⁽¹⁾

	MIN	MAX	UNIT
Supply voltage		±18	V
Analog input voltage		±40	V
Output short circuit (to ground)		continuous	
Operating temperature	-40	125	°C
Junction temperature		150	°C
Lead temperature (soldering, 10 seconds)		300	°C
Storage temperature, T _{stg}	-55	125	°C

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

7.2 ESD Ratings

		VALUE	UNIT
V _(ESD)	Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	±2000
		Charged-device model (CDM), per JEDEC specification JESD22-C101 ⁽²⁾	±50

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

7.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

	MIN	NOM	MAX	UNIT
V power supply	±2.25	±15	±18	V
Input common-mode voltage range for $V_O = 0$	$V - 2\text{ V}$		$V + -2\text{ V}$	
T_A operating temperature INA128-HT	-55		175	°C
T_A operating temperature INA129-HT	-55		210	°C

7.4 Thermal Information

THERMAL METRIC ⁽¹⁾		INA12x		UNIT
		D (SOIC)	P (PDIP)	
		8 PINS	8 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	110	46.1	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	57	34.1	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	54	23.4	°C/W
Ψ_{JT}	Junction-to-top characterization parameter	11	11.3	°C/W
Ψ_{JB}	Junction-to-board characterization parameter	53	23.2	°C/W

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics application report](#).

7.5 Electrical Characteristics

at $T_A = 25^\circ\text{C}$, $V_S = \pm 15\text{ V}$, and $R_L = 10\text{ k}\Omega$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
INPUT							
Offset voltage, RTI	Initial	$T_A = 25^\circ\text{C}$	INA128P, U INA129P, U	±10±100/G	±50±500/G		μV
			INA128PA, UA INA129PA, UA	±25±100/G	±125±1000/G		
	vs temperature	$T_A = T_{MIN}$ to T_{MAX}	INA128P, U INA129P, U	±0.2±2/G	±0.5±20/G		μV/°C
			INA128PA, UA INA129PA, UA	±0.2±5/G	±1±20/G		
	vs power supply	$V_S = \pm 2.25\text{ V}$ to $\pm 18\text{ V}$	INA128P, U INA129P, U	±0.2±20/G	±1±100/G		μV/V
INA128PA, UA INA129PA, UA			±2±200/G				
Long-term stability			±0.1±3/g			μV/mo	
Impedance	Differential			$10^{10} \parallel 2$			Ω pF
	Common mode			$10^{11} \parallel 9$			
Common-mode voltage range ⁽¹⁾		$V_O = 0\text{ V}$		(V+) - 2	(V+) - 1.4		V
				(V...) + 2	(V-) + 1.7		
Safe input voltage						±40	V

(1) Input common-mode range varies with output voltage; see [Typical Characteristics](#).

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Electrical Characteristics (continued)

at $T_A = 25^\circ\text{C}$, $V_S = \pm 15\text{ V}$, and $R_L = 10\text{ k}\Omega$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
Common-mode rejection		$V_{CM} = \pm 13\text{ V}$, $\Delta R_S = 1\text{ k}\Omega$	G = 1	INA128P, U INA129P, U	80	86	dB
				INA128PA, UA INA129PA, UA	73		
			G = 10	INA128P, U INA129P, U	100	106	
				INA128PA, UA INA129PA, UA	93		
		G = 100	INA128P, U INA129P, U	120	125		
			INA128PA, UA INA129PA, UA	110			
		G = 1000	INA128P, U INA129P, U	120	130		
			INA128PA, UA INA129PA, UA	110			
Bias current		INA128P, U INA129P, U		± 2	± 5	nA	
		INA128PA, UA INA129PA, UA			± 10		
Bias current vs temperature				± 30		pA/°C	
Offset current		INA128P, U INA129P, U		± 1	± 5	nA	
		INA128PA, UA INA129PA, UA			± 10		
Offset current vs temperature				± 30		pA/°C	
Noise voltage, RTI	f = 10 Hz	f = 100 Hz	f = 1 kHz	$f_B = 0.1\text{ Hz to }10\text{ Hz}$	G = 1000, $R_S = 0\Omega$	10	nV/√Hz
						8	
						8	
						0.2	
Noise current	f = 10 Hz	f = 1 kHz	$F_B = 0.1\text{ Hz to }10\text{ Hz}$		0.9	pA/√Hz	
					0.3		
					30		
GAIN⁽²⁾							
Gain equation	INA128			$1 + (50\text{ k}\Omega/R_G)$		V/V	
	INA129			$1 + (49.4\text{ k}\Omega/R_G)$			
Range of gain				1		10000	V/V
Gain error	G = 1		INA128P, U INA129P, U	$\pm 0.01\%$	$\pm 0.024\%$		
			INA128PA, UA INA129PA, UA		$\pm 0.01\%$		
	G = 10		INA128P, U INA129P, U	$\pm 0.02\%$	$\pm 0.4\%$		
			INA128PA, UA INA129PA, UA		$\pm 0.5\%$		
	G = 100		INA128P, U INA129P, U	$\pm 0.05\%$	$\pm 0.5\%$		
			INA128PA, UA INA129PA, UA		$\pm 0.7\%$		
	G = 1000		INA128P, U INA129P, U	$\pm 0.5\%$	$\pm 1\%$		
			INA128PA, UA INA129PA, UA		$\pm 2\%$		

(2) Nonlinearity measurements in G = 1000 are dominated by noise. Typical non-linearity is $\pm 0.001\%$.

Electrical Characteristics (continued)

 at $T_A = 25^\circ\text{C}$, $V_S = \pm 15\text{ V}$, and $R_L = 10\text{ k}\Omega$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
Gain vs temperature ⁽³⁾	G = 1			±1	±10	ppm/°C
	50-k Ω (or 49.4-k Ω) Resistance ⁽³⁾⁽⁴⁾			±25	±100	
Nonlinearity	$V_O = \pm 13.6\text{ V}$, G = 1	INA128P, U INA129P, U		±0.0001	±0.001	% of FSR
		INA128PA, UA INA129PA, UA			±0.002	
	G = 10	INA128P, U INA129P, U		±0.0003	±0.002	
		INA128PA, UA INA129PA, UA			±0.004	
	G = 100	INA128P, U INA129P, U		±0.0005	±0.002	
		INA128PA, UA INA129PA, UA			±0.004	
G = 1000			±0.001	/>		
OUTPUT⁽²⁾						
Voltage	Positive	$R_L = 10\text{ k}\Omega$	(V+) – 1.4	(V+) – 0.9		V
	Negative	$R_L = 10\text{ k}\Omega$	(V-) + 1.4	(V-) + 0.8		
Load capacitance stability				1000		pF
Short-circuit current				6/–15		mA
FREQUENCY RESPONSE						
Bandwidth, –3 dB	G = 1			1.3		MHz
	G = 10			700		kHz
	G = 100			200		
	G = 1000			20		
Slew rate	$V_O = \pm 10\text{ V}$, G = 10			4		V/ μs
Settling time, 0.01%	G = 1			7		μs
	G = 10			7		
	G = 100			9		
	G = 1000			80		
Overload recovery	50% overdrive			4		μs
POWER SUPPLY						
Voltage range			±2.25	±15	±18	V
Current, total	$V_{IN} = 0\text{ V}$			±700	±750	μA
TEMPERATURE RANGE						
Specification			–40		85	°C
Operating			–40		125	°C

(3) Specified by wafer test.

 (4) Temperature coefficient of the 50 k Ω (or 49.4 k Ω) term in the gain equation.

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7.6 Typical Characteristics

at $T_A = 25^\circ\text{C}$ and $V_S = \pm 15\text{ V}$ (unless otherwise noted)

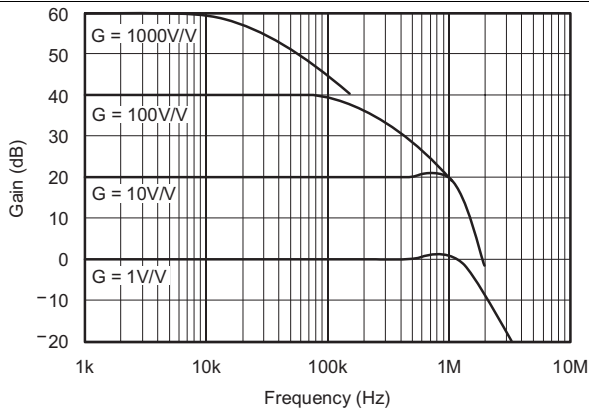


Figure 1. Gain vs Frequency

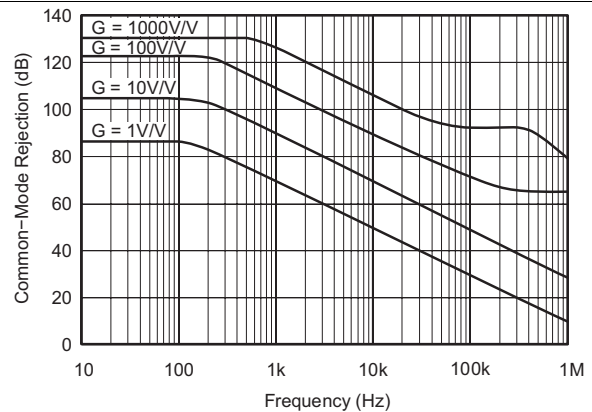


Figure 2. Common-Mode Rejection vs Frequency

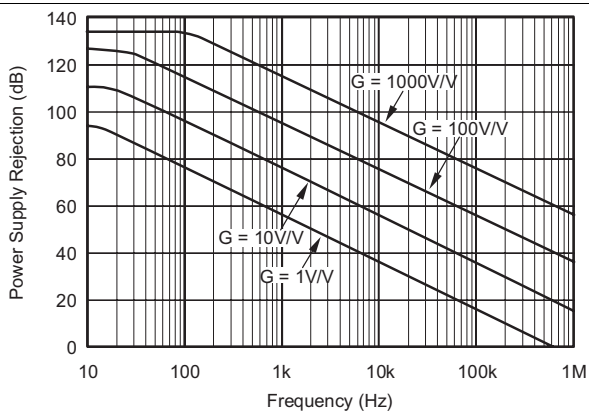


Figure 3. Positive Power Supply Rejection vs Frequency

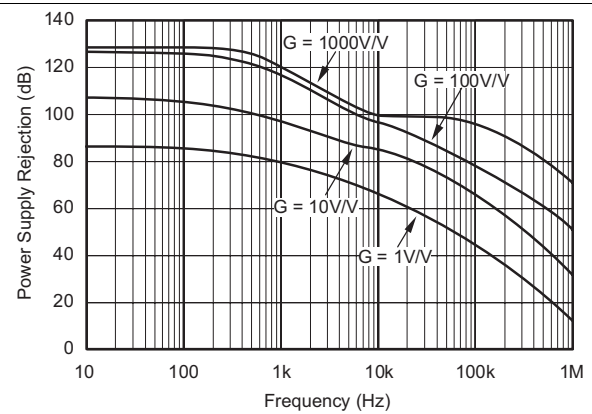


Figure 4. Negative Power Supply Rejection vs Frequency

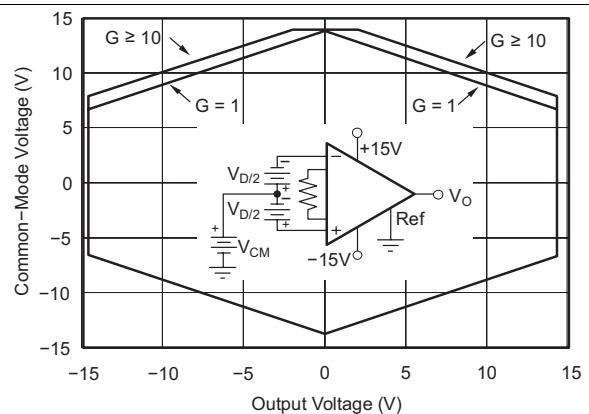


Figure 5. Input Common-Mode Range vs Output Voltage, $V_S = \pm 15\text{ V}$

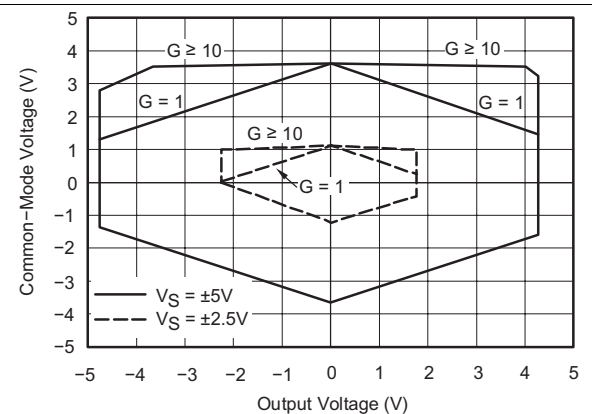


Figure 6. Input Common-Mode Range vs Output Voltage, $V_S = \pm 5\text{ V}, \pm 2.5\text{ V}$

Typical Characteristics (continued)

at $T_A = 25^\circ\text{C}$ and $V_S = \pm 15\text{ V}$ (unless otherwise noted)

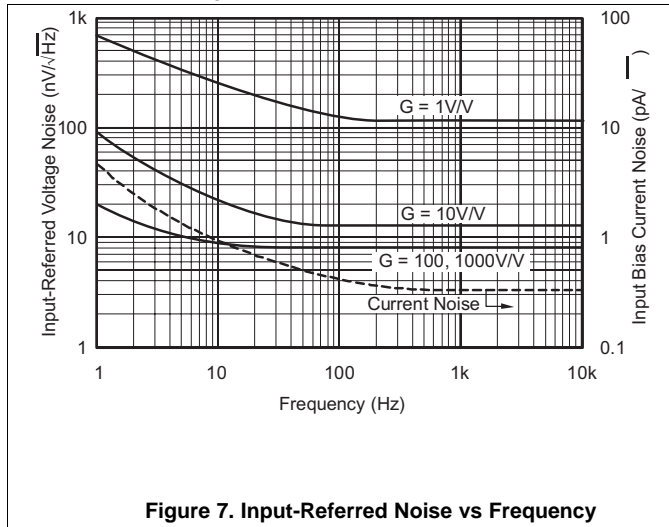


Figure 7. Input-Referred Noise vs Frequency

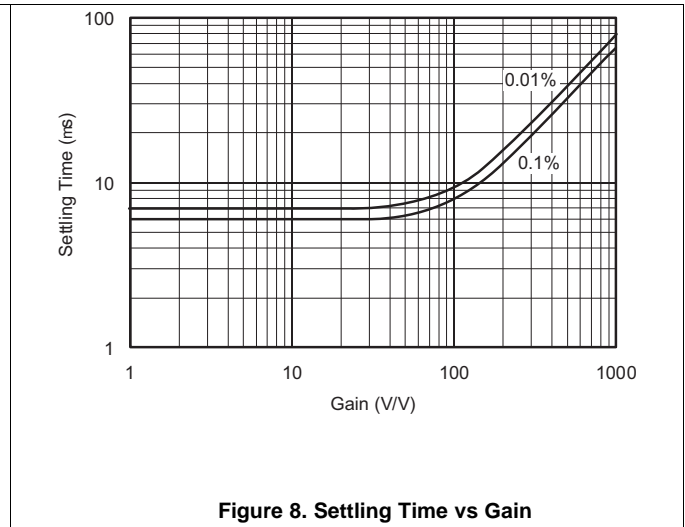


Figure 8. Settling Time vs Gain

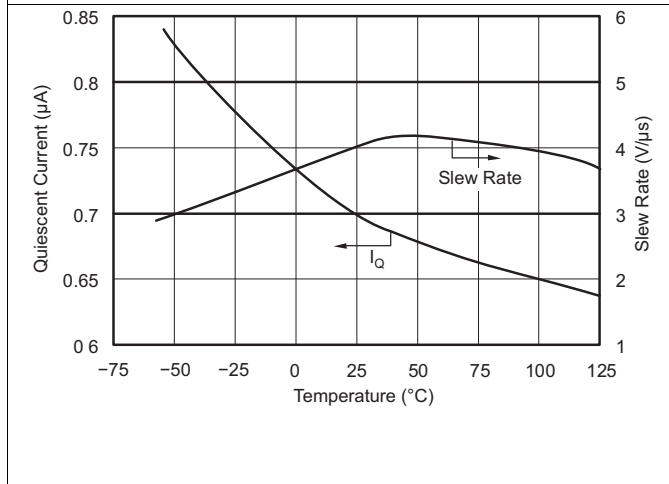


Figure 9. Quiescent Current and Slew Rate vs Temperature

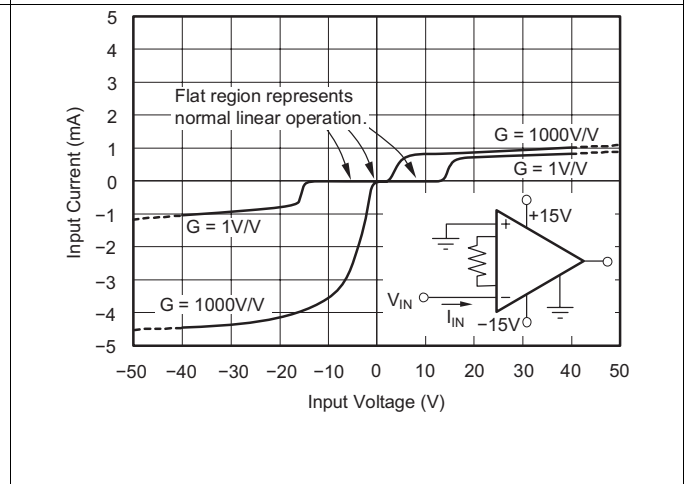


Figure 10. Input Overvoltage V_I Characteristics

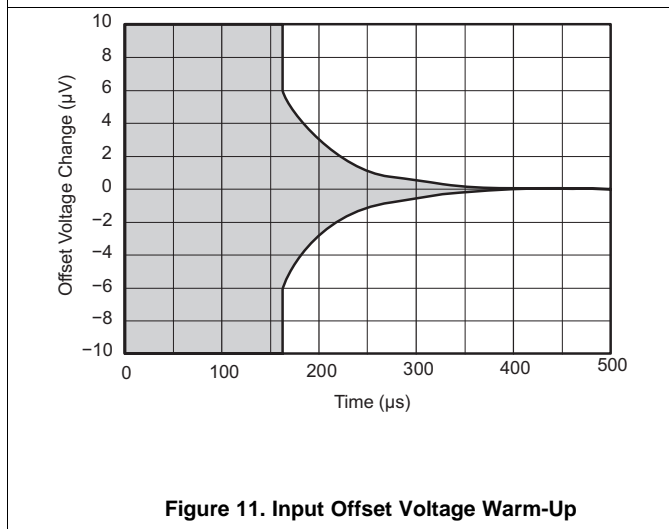


Figure 11. Input Offset Voltage Warm-Up

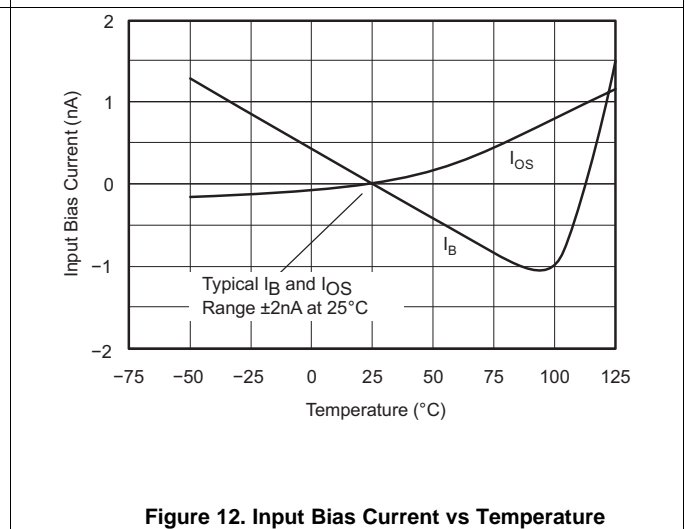


Figure 12. Input Bias Current vs Temperature

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Typical Characteristics (continued)

at $T_A = 25^\circ\text{C}$ and $V_S = \pm 15\text{ V}$ (unless otherwise noted)

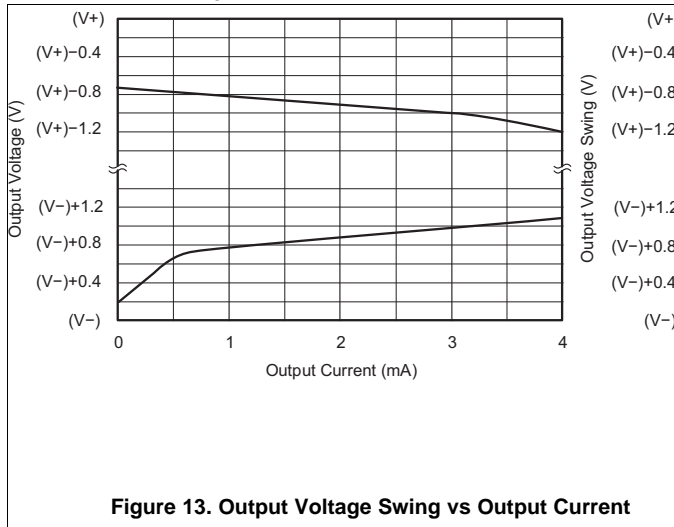


Figure 13. Output Voltage Swing vs Output Current

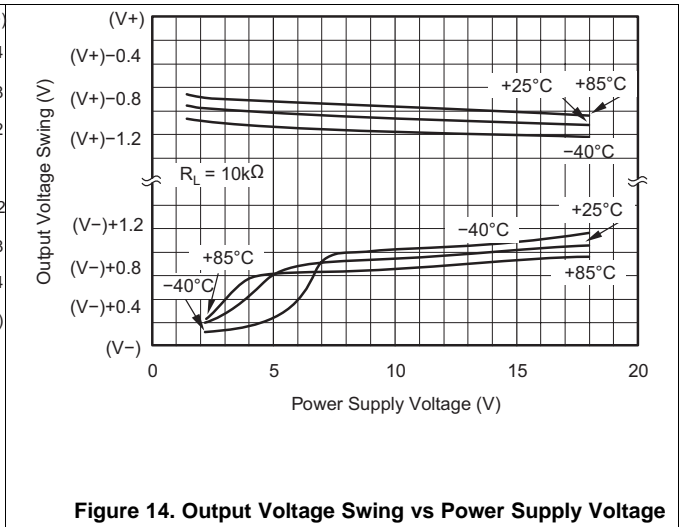


Figure 14. Output Voltage Swing vs Power Supply Voltage

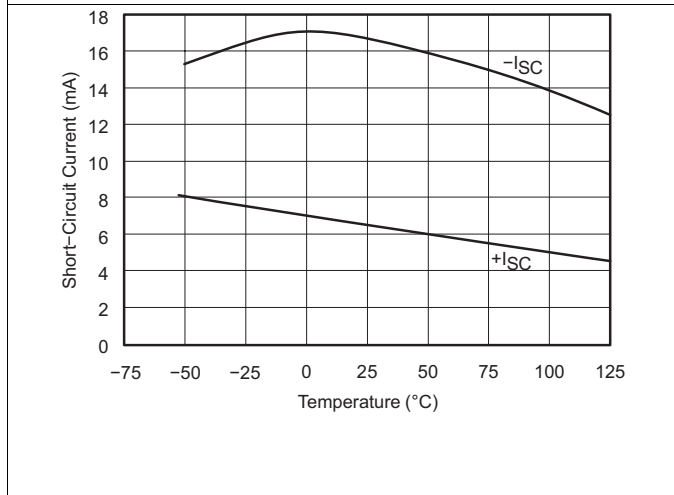


Figure 15. Short Circuit Output Current vs Temperature

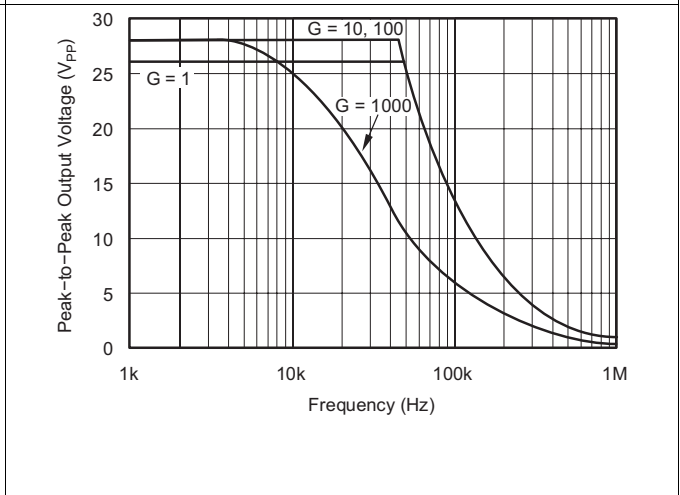


Figure 16. Maximum Output Voltage vs Frequency

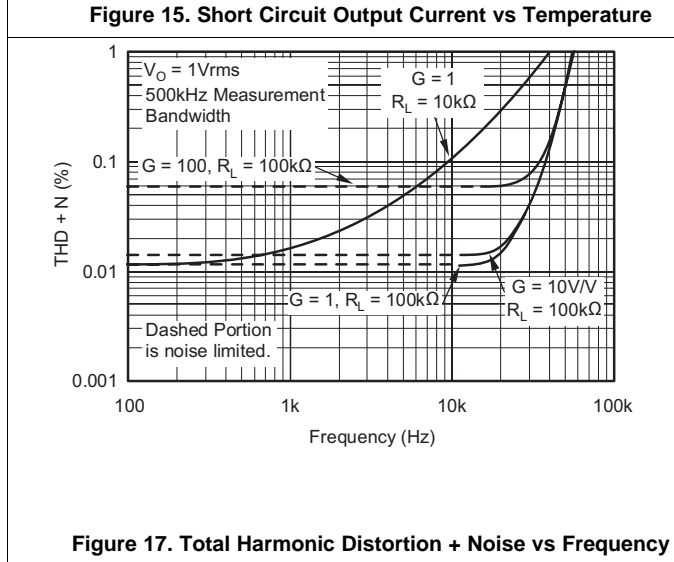


Figure 17. Total Harmonic Distortion + Noise vs Frequency

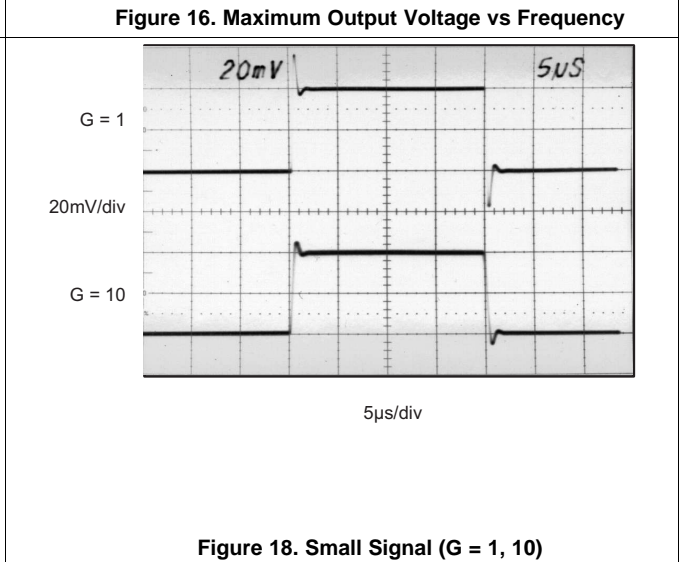


Figure 18. Small Signal (G = 1, 10)

Typical Characteristics (continued)

at $T_A = 25^\circ\text{C}$ and $V_S = \pm 15\text{ V}$ (unless otherwise noted)

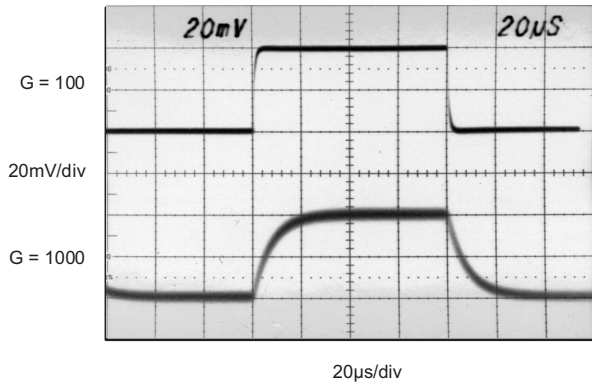


Figure 19. Small Signal (G = 100, 1000)

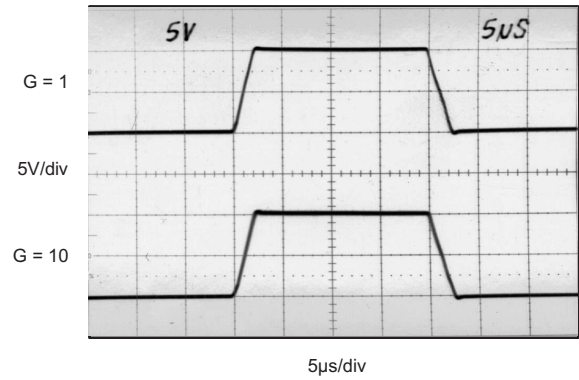


Figure 20. Large Signal (G = 1, 10)

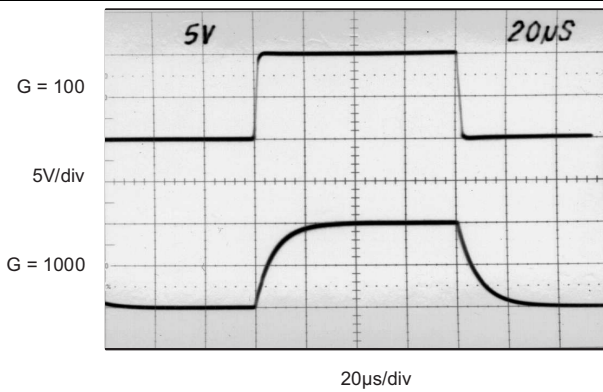


Figure 21. Large Signal (G = 100, 1000)

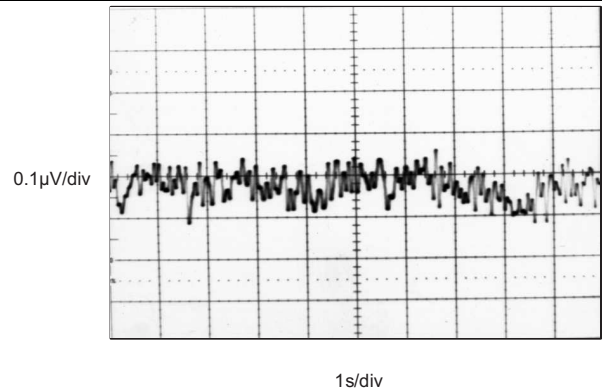


Figure 22. Voltage Noise 0.1 to 10-Hz Input-Referred, $G \geq 100$

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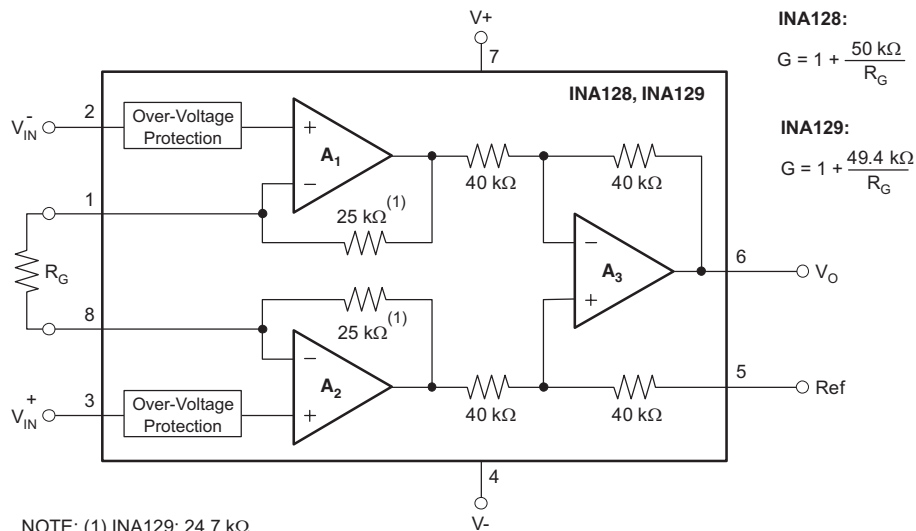
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8 Detailed Description

8.1 Overview

The INA12x instrumentation amplifier is a type of differential amplifier that has been outfitted with input protection circuit and input buffer amplifiers, which eliminate the need for input impedance matching and make the amplifier particularly suitable for use in measurement and test equipment. Additional characteristics of the INA128 include a very low DC offset, low drift, low noise, very high open-loop gain, very high common-mode rejection ratio, and very high input impedances. The INA12x is used where great accuracy and stability of the circuit both short and long term are required.

8.2 Functional Block Diagram



8.3 Feature Description

The INA12x devices are low power, general-purpose instrumentation amplifiers offering excellent accuracy. The versatile three-operational-amplifier design and small size make the amplifiers ideal for a wide range of applications. Current-feedback input circuitry provides wide bandwidth, even at high gain. A single external resistor sets any gain from 1 to 10,000. The INA128 is laser trimmed for very low offset voltage (25 μ V typical) and high common-mode rejection (93 dB at $G \geq 100$). These devices operate with power supplies as low as ± 2.25 V, and quiescent current of 2 mA, typically. The internal input protection can withstand up to ± 40 V without damage.

8.4 Device Functional Modes

8.4.1 Noise Performance

The INA12x provides very low noise in most applications. Low-frequency noise is approximately $0.2 \mu\text{V}_{\text{PP}}$ measured from 0.1 to 10 Hz ($G \geq 100$). This provides dramatically improved noise when compared to state-of-the-art chopper-stabilized amplifiers.

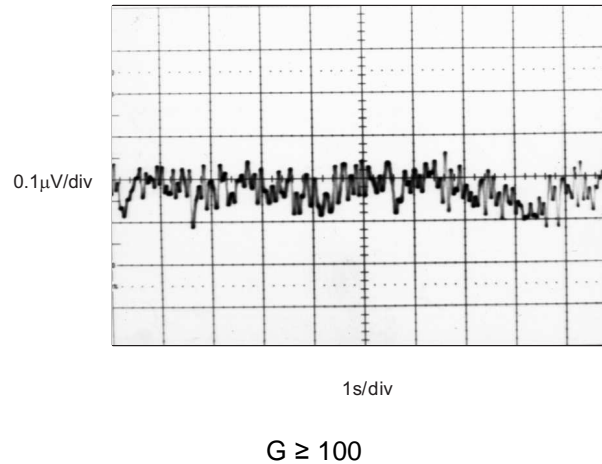


Figure 23. 0.1-Hz to 10-Hz Input-Referred Voltage Noise

8.4.2 Input Common-Mode Range

The linear input voltage range of the input circuitry of the INA12x is from approximately 1.4 V below the positive supply voltage to 1.7 V above the negative supply. As a differential input voltage causes the output voltage increase, however, the linear input range is limited by the output voltage swing of amplifiers A_1 and A_2 . Thus the linear common-mode input range is related to the output voltage of the complete amplifier. This behavior also depends on supply voltage (see performance curve [Figure 6](#)).

Input-overload can produce an output voltage that appears normal. For example, if an input overload condition drives both input amplifiers to their positive output swing limit, the difference voltage measured by the output amplifier will be near zero. The output of A_3 will be near 0 V even though both inputs are overloaded.

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9 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 Application Information

The INA12x measures small differential voltage with high common-mode voltage developed between the noninverting and inverting input. The high-input voltage protection circuit in conjunction with high input impedance make the INA12x suitable for a wide range of applications. The ability to set the reference pin to adjust the functionality of the output signal offers additional flexibility that is practical for multiple configurations.

9.2 Typical Application

Figure 24 shows the basic connections required for operation of the INA12x. Applications with noisy or high impedance power supplies may require decoupling capacitors close to the device pins as shown. The output is referred to the output reference (Ref) terminal which is normally grounded. This must be a low-impedance connection to assure good common-mode rejection. A resistance of 8 Ω in series with the Ref pin will cause a typical device to degrade to approximately 80dB CMR (G = 1).

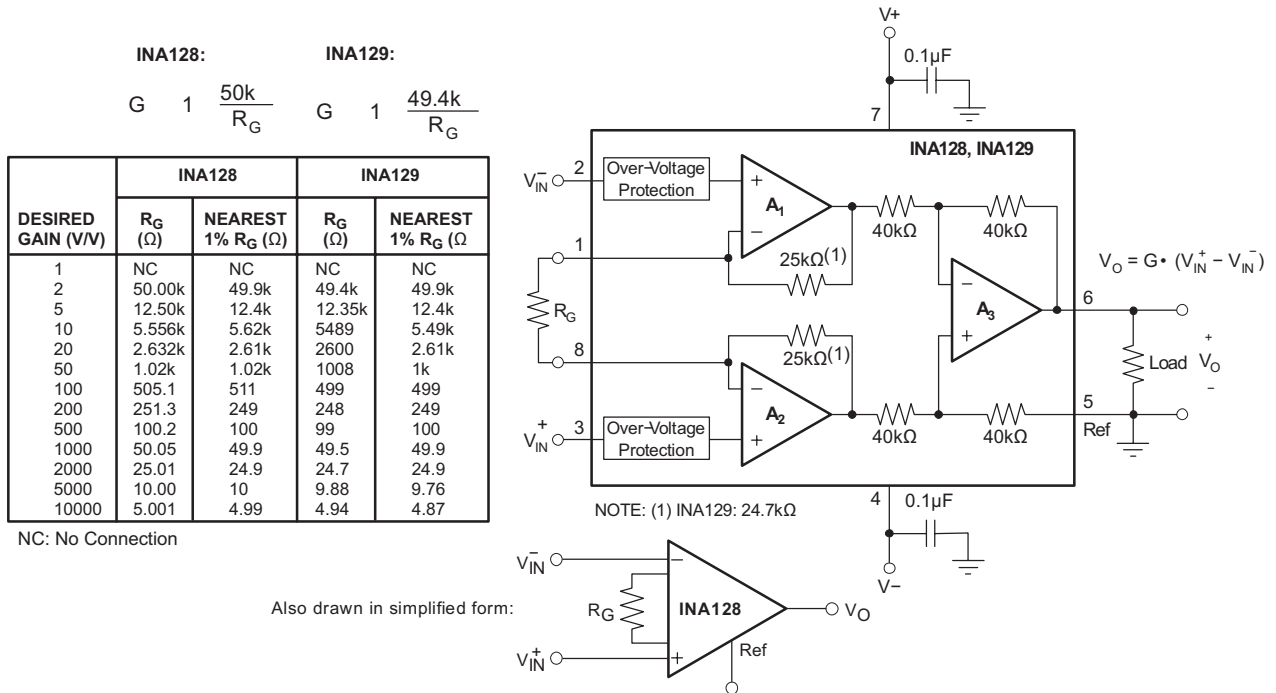


Figure 24. Basic Connections

Typical Application (continued)

9.2.1 Design Requirements

The device can be configured to monitor the input differential voltage when the gain of the input signal is set by the external resistor R_G . The output signal references to the Ref pin. The most common application is where the output is referenced to ground when no input signal is present by connecting the Ref pin to ground, as Figure 24 shows. When the input signal increases, the output voltage at the OUT pin increases, too.

9.2.2 Detailed Design Procedure

9.2.2.1 Setting the Gain

Gain is set by connecting a single external resistor, R_G , connected between pins 1 and 8:

$$\text{INA128: } g = 1 + 50 \text{ k}\Omega/R_G \quad (1)$$

Commonly used gains and resistor values are shown in Figure 24.

The 50-k Ω term in Equation 1 comes from the sum of the two internal feedback resistors of A_1 and A_2 . These on-chip metal film resistors are laser-trimmed to accurate absolute values. The accuracy and temperature coefficient of these internal resistors are included in the gain accuracy and drift specifications of the INA128.

The stability and temperature drift of the external gain setting resistor, R_G , also affects gain. The contribution of R_G to gain accuracy and drift can be directly inferred from Equation 1. Low resistor values required for high gain can make wiring resistance important. Sockets add to the wiring resistance, which contributes additional gain error (possibly an unstable gain error) in gains of approximately 100 or greater.

9.2.2.2 Dynamic Performance

The typical performance curve Figure 1 shows that, despite its low quiescent current, the INA12x achieves wide bandwidth even at high gain. This is due to the current-feedback topology of the input stage circuitry. Settling time also remains excellent at high gain.

9.2.2.3 Offset Trimming

The INA12x is laser-trimmed for low-offset voltage and offset voltage drift. Most applications require no external offset adjustment. Figure 25 shows an optional circuit for trimming the output offset voltage. The voltage applied to the Ref terminal is summed with the output. The op amp buffer provides low impedance at the Ref terminal to preserve good common-mode rejection.

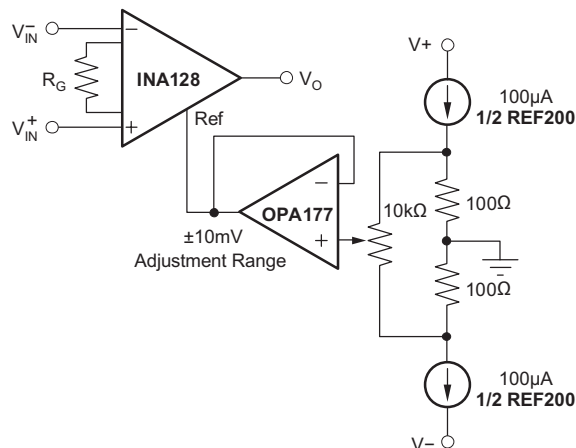


Figure 25. Optional Trimming of Output Offset Voltage

9.2.2.4 Input Bias Current Return Path

The input impedance of the INA12x is extremely high: approximately $10^{10} \Omega$. However, a path must be provided for the input bias current of both inputs. This input bias current is approximately ± 2 nA. High input impedance means that this input bias current changes very little with varying input voltage.

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Typical Application (continued)

Input circuitry must provide a path for this input bias current for proper operation. Figure 26 shows various provisions for an input bias current path. Without a bias current path, the inputs will float to a potential which exceeds the common-mode range, and the input amplifiers will saturate.

If the differential source resistance is low, the bias current return path can be connected to one input (see the thermocouple example in Figure 26). With higher source impedance, using two equal resistors provides a balanced input, with possible advantages of lower input offset voltage due to bias current and better high-frequency common-mode rejection.

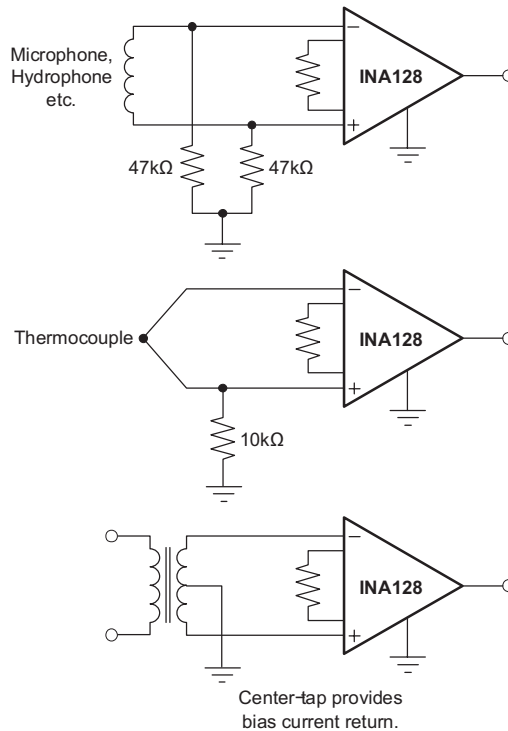
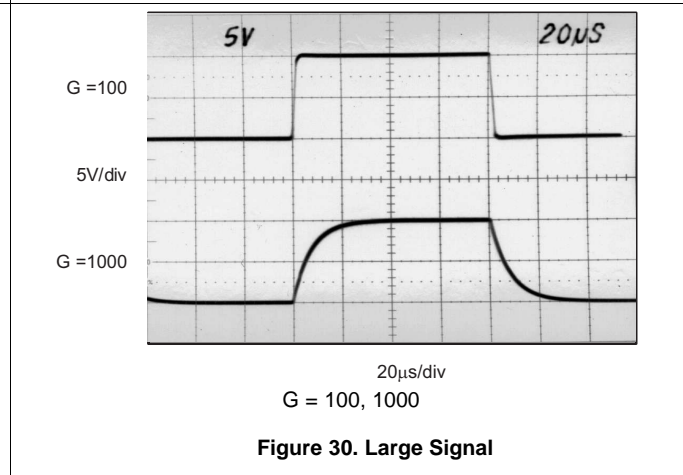
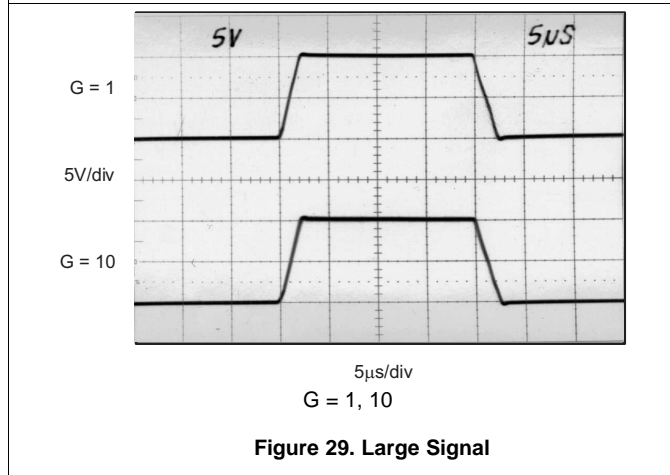
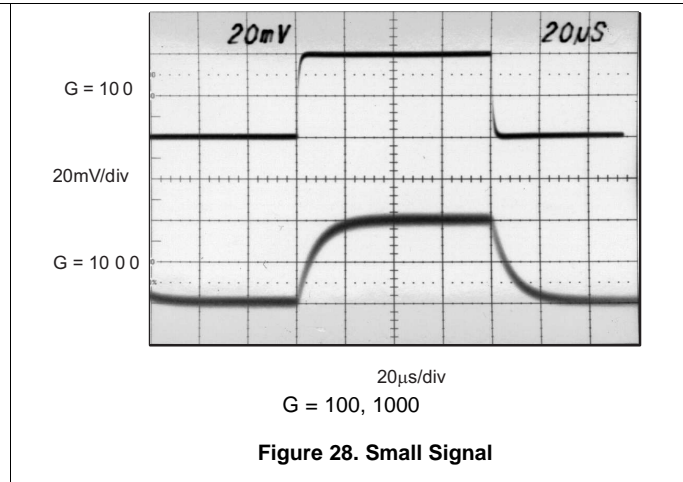
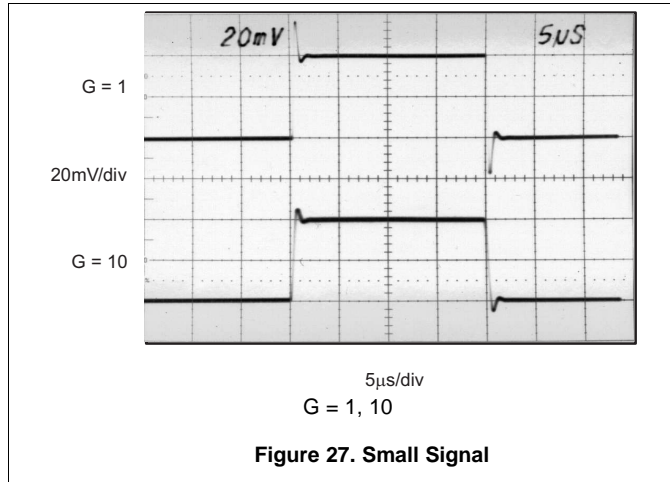


Figure 26. Providing an Input Common-Mode Current Path

Typical Application (continued)

9.2.3 Application Curves



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10 Power Supply Recommendations

The minimum power supply voltage for INA12x is ± 2.25 V and the maximum power supply voltage is ± 18 V. This minimum and maximum range covers a wide range of power supplies; but for optimum performance, ± 15 V is recommended. TI recommends adding a bypass capacitor at the input to compensate for the layout and power supply source impedance.

10.1 Low Voltage Operation

The INA12x can be operated on power supplies as low as ± 2.25 V. Performance remains excellent with power supplies ranging from ± 2.25 V to ± 18 V. Most parameters vary only slightly throughout this supply voltage range—see [Typical Characteristics](#).

Operation at very low supply voltage requires careful attention to assure that the input voltages remain within their linear range. Voltage swing requirements of internal nodes limit the input common-mode range with low power supply voltage. [Figure 6](#) shows the range of linear operation for ± 15 -V, ± 5 -V, and ± 2.5 -V supplies.

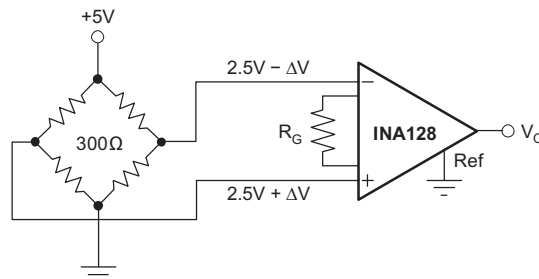


Figure 31. Bridge Amplifier

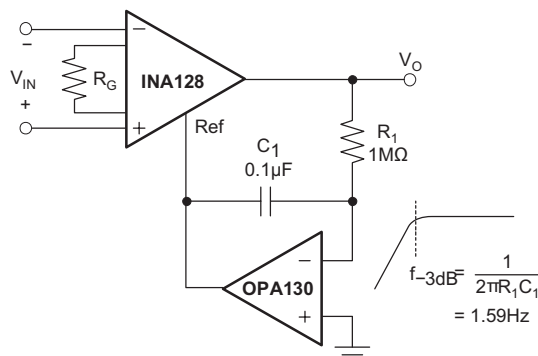
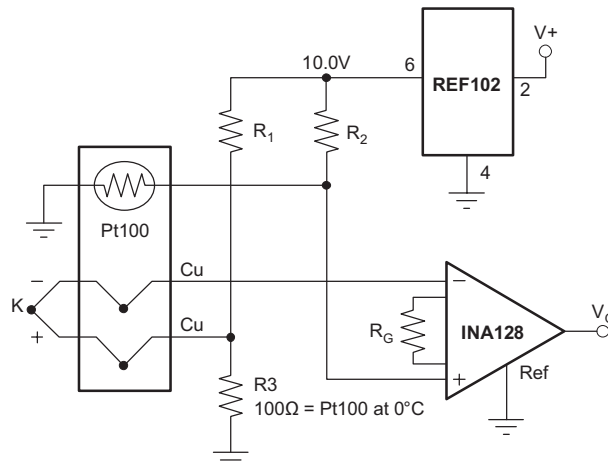


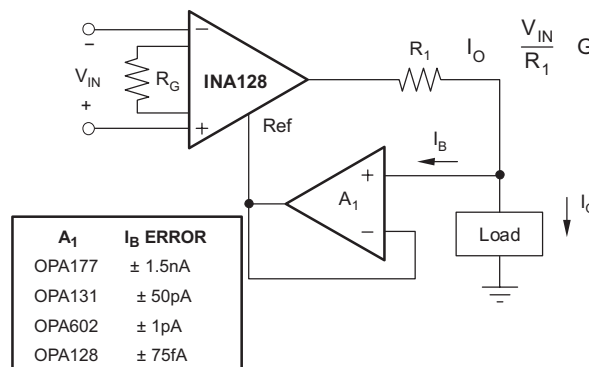
Figure 32. AC-Coupled Instrumentation Amplifier

Low Voltage Operation (continued)



ISA TYPE	MATERIAL	SEEBECK COEFFICIENT (C)	R ₁ , R ₂
E	+ Chromel - Constantan	58.5	66.5kΩ
J	+ Iron - Constantan	50.2	76.8kΩ
K	+ Chromel - Alumel	39.4	97.6kΩ
T	+ Copper - Constantan	38.0	102kΩ

Figure 33. Thermocouple Amplifier With RTD Cold-Junction Compensation



A ₁	I _B ERROR
OPA177	± 1.5nA
OPA131	± 50pA
OPA602	± 1pA
OPA128	± 75fA

Figure 34. Differential Voltage to Current Converter

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Low Voltage Operation (continued)

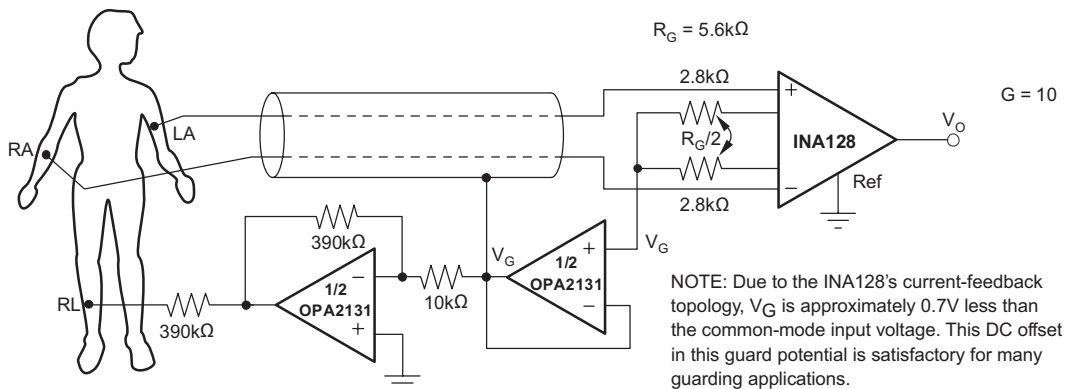


Figure 35. ECG Amplifier With Right-Leg Drive

11 Layout

11.1 Layout Guidelines

Place the power-supply bypass capacitor as closely as possible to the supply and ground pins. The recommended value of this bypass capacitor is 0.1 μF to 1 μF. If necessary, additional decoupling capacitance can be added to compensate for noisy or high-impedance power supplies. These decoupling capacitors must be placed between the power supply and INA12x devices.

The gain resistor must be placed close to pin 1 and pin 8. This placement limits the layout loop and minimizes any noise coupling into the part.

11.2 Layout Example

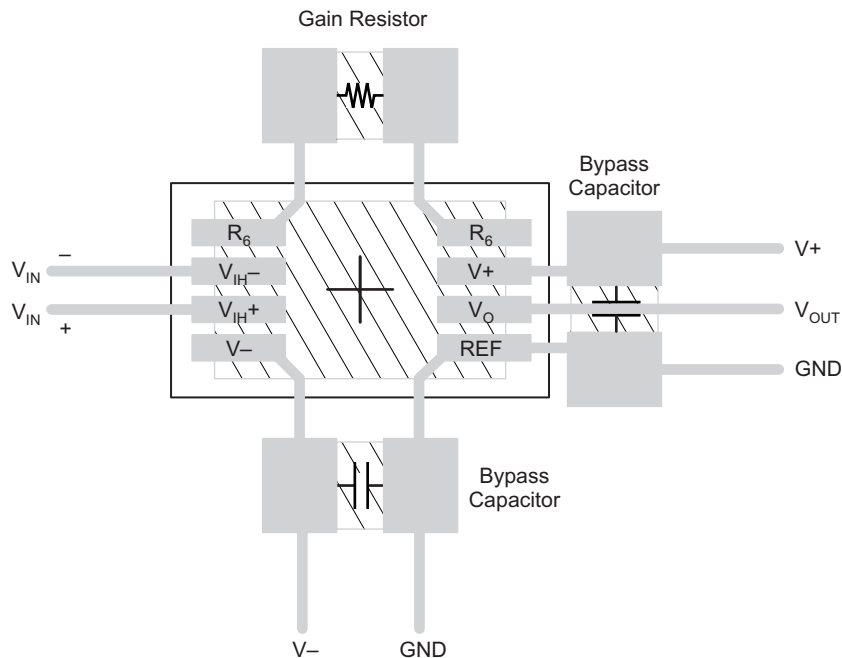


Figure 36. Recommended Layout

12 Device and Documentation Support

12.1 Related Links

Table 1 lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to sample or buy.

Table 1. Related Links

PARTS	PRODUCT FOLDER	SAMPLE & BUY	TECHNICAL DOCUMENTS	TOOLS & SOFTWARE	SUPPORT & COMMUNITY
INA128	Click here	Click here	Click here	Click here	Click here
INA129	Click here	Click here	Click here	Click here	Click here

12.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

12.3 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of Use](#).

TI E2E™ Online Community *TI's Engineer-to-Engineer (E2E) Community*. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

Design Support *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

12.4 Trademarks

E2E is a trademark of Texas Instruments.
All other trademarks are the property of their respective owners.

12.5 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

12.6 Glossary

SLYZ022 — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	O
INA128P	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	
INA128PA	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	
INA128PG4	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	
INA128U	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-3-260C-168 HR	
INA128U/2K5	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-3-260C-168 HR	
INA128U/2K5G4	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-3-260C-168 HR	
INA128UA	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-3-260C-168 HR	
INA128UA/2K5	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-3-260C-168 HR	
INA128UA/2K5E4	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-3-260C-168 HR	
INA128UA/2K5G4	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-3-260C-168 HR	
INA128UAE4	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-3-260C-168 HR	
INA128UAG4	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-3-260C-168 HR	
INA128UG4	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-3-260C-168 HR	
INA129P	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	
INA129PA	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	

⁽⁵⁾ Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line of the previous line and the two combined represent the entire Device Marking for that device.

⁽⁶⁾ Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/End lines if the finish value exceeds the maximum column width.

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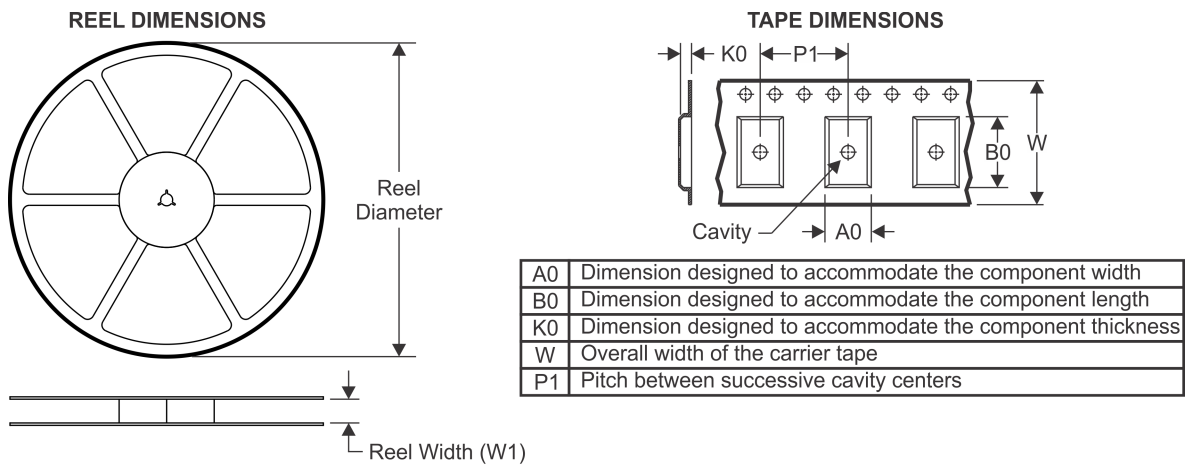
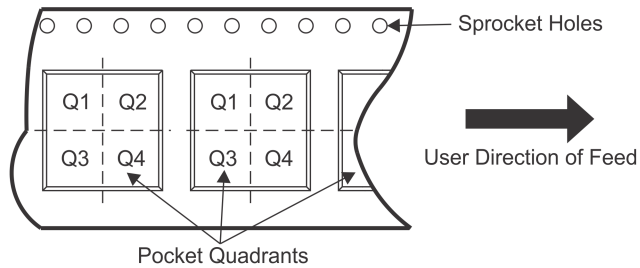
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OTHER QUALIFIED VERSIONS OF INA128, INA129 :

- Enhanced Product: [INA129-EP](#)

NOTE: Qualified Version Definitions:

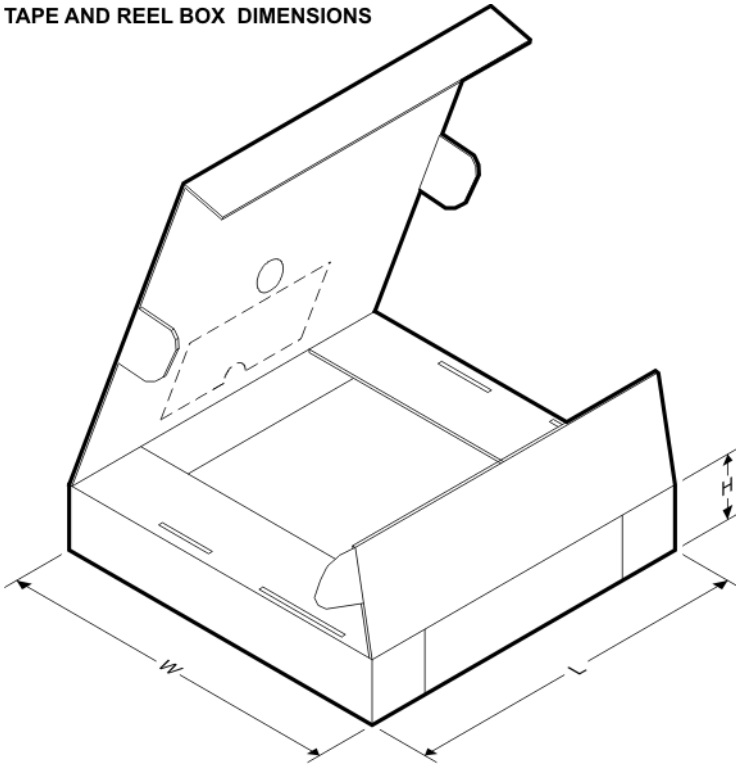
- Enhanced Product - Supports Defense, Aerospace and Medical Applications

TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE


*All dimensions are nominal

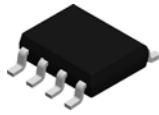
Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
INA128U/2K5	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
INA128UA/2K5	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
INA129U/2K5	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
INA129UA/2K5	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
INA128U/2K5	SOIC	D	8	2500	853.0	449.0	35.0
INA128UA/2K5	SOIC	D	8	2500	853.0	449.0	35.0
INA129U/2K5	SOIC	D	8	2500	367.0	367.0	35.0
INA129UA/2K5	SOIC	D	8	2500	853.0	449.0	35.0

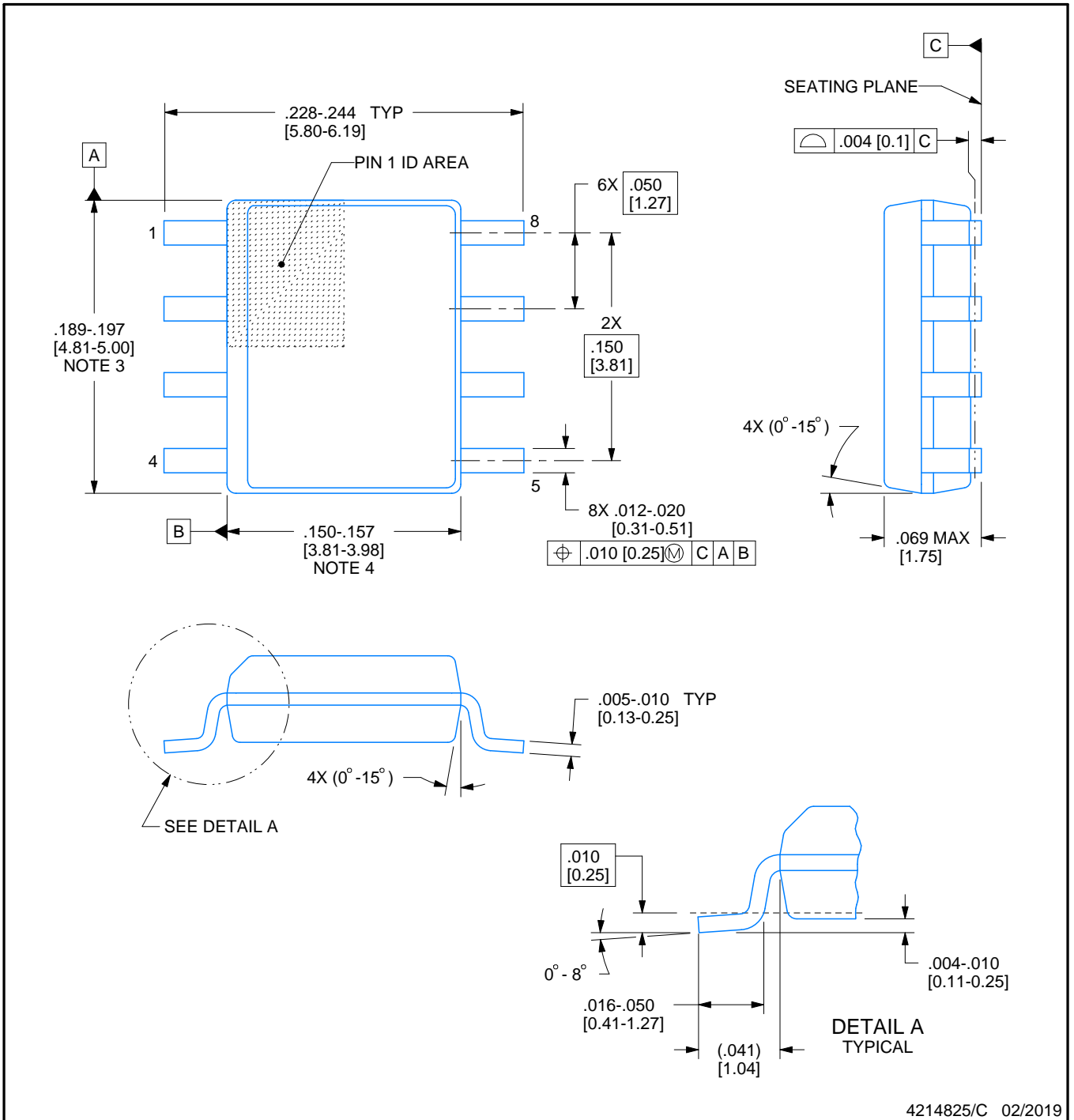


D0008A

PACKAGE OUTLINE

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



4214825/C 02/2019

NOTES:

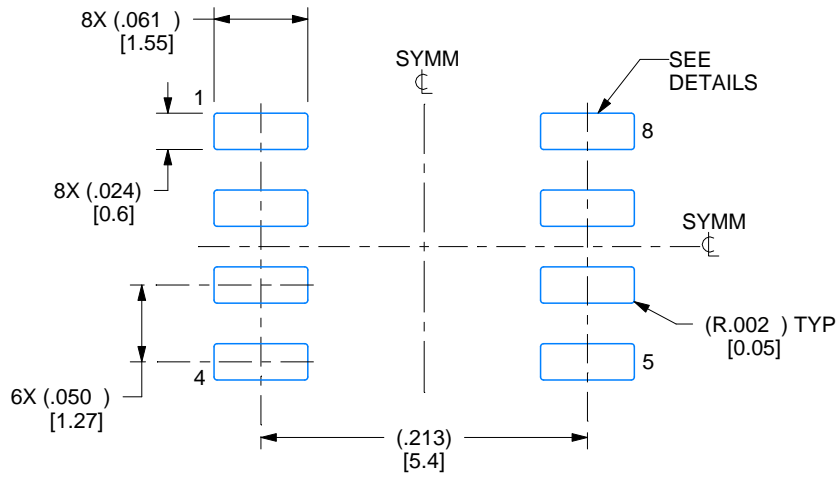
- Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
- This drawing is subject to change without notice.
- This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed $.006$ [0.15] per side.
- This dimension does not include interlead flash.
- Reference JEDEC registration MS-012, variation AA.

EXAMPLE BOARD LAYOUT

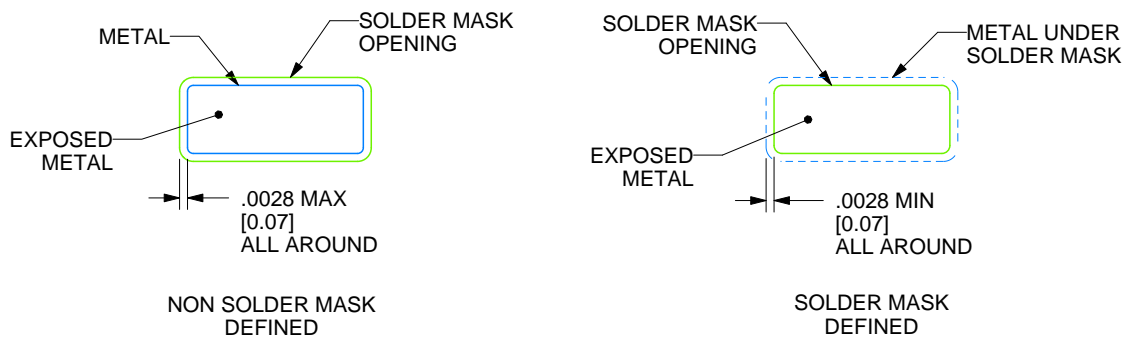
D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE:8X



SOLDER MASK DETAILS

4214825/C 02/2019

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

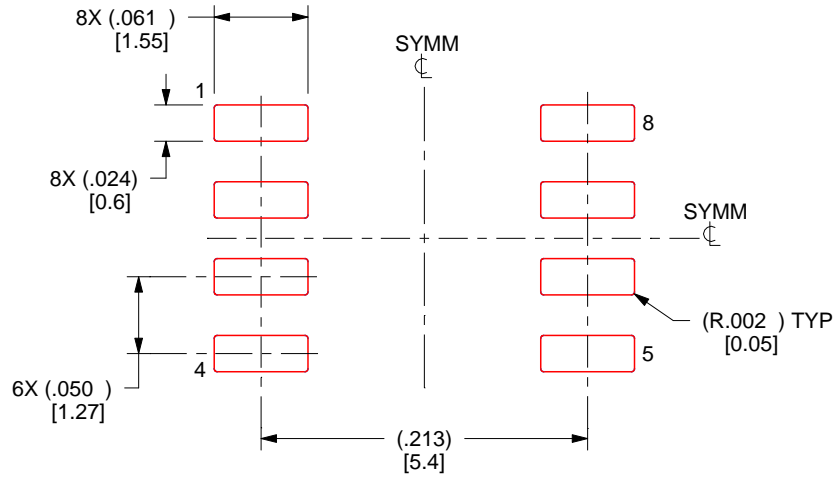
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



SOLDER PASTE EXAMPLE
BASED ON .005 INCH [0.125 MM] THICK STENCIL
SCALE:8X

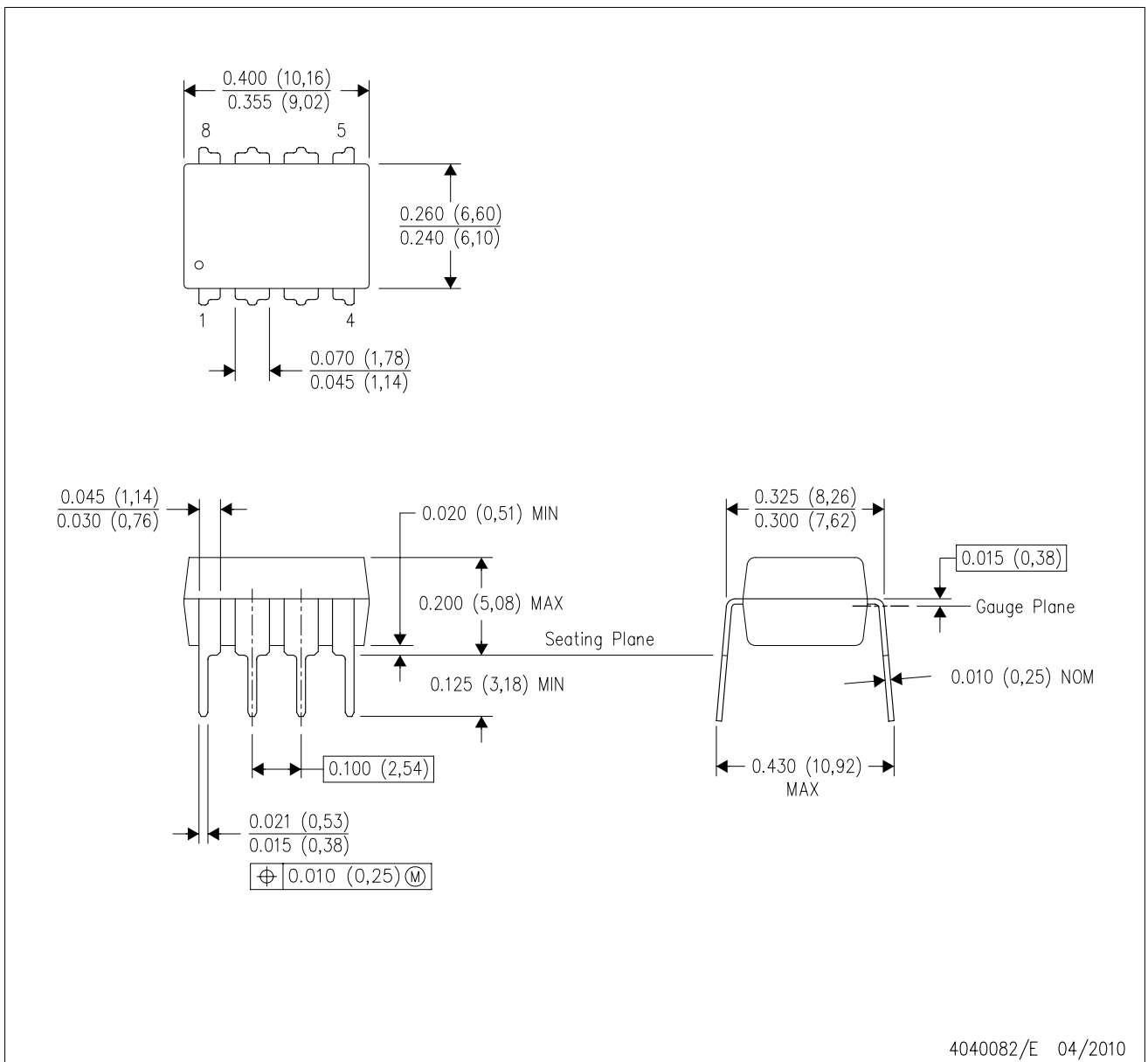
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NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

P (R-PDIP-T8)

PLASTIC DUAL-IN-LINE PACKAGE



4040082/E 04/2010

- NOTES:
- A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - C. Falls within JEDEC MS-001 variation BA.

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